

The BusQor[®] SQ60120EPB25 bus converter is a nextgeneration, board-mountable, isolated, fixed switching frequency DC-DC converter that uses synchronous rectification to achieve extremely high conversion efficiency. The power dissipated by the converter is so low that a heatsink is not required, which saves cost, weight, height, and application effort. The SQ BusQor series provides an isolated step down voltage from 48V to a semi-regulated 12V intermediate bus. BusQor converters are ideal for customers who need multiple outputs and wish to build or buy point of load converters to work with a 12V rail. RoHS Compliant (see last page).



Operational Features

- High efficiency, 95% at full rated load current
- Delivers 25A (300W) power with minimal derating - no heatsink required
- Operating input voltage range: 36-75V
- Fixed frequency switching provides predictable EMI
- No minimum load requirement

Control Features

- On/Off control referenced to input side
- Inherent current share (by droop method) for high current and parallel applications.

Mechanical Features

- Industry standard quarter-brick pin-out configuration
- Size: 2.30" x 0.96" (58.4 x 24.4mm), 0.563" (14.30mm)
- Total weight: 1.62 oz (46.2 g)

Protection Features

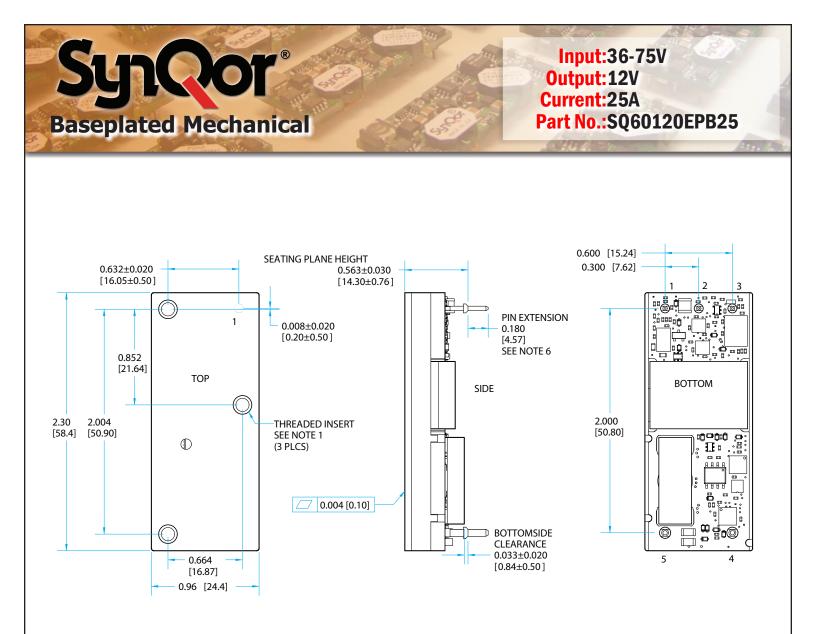
- Input under-voltage lockout and over-voltage shutdown protects against abnormal input voltages
- Output current limit and short circuit protection
- Output over-voltage protection
- Thermal shutdown

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Safety Features

- CAN/CSA C22.2 No. 60950-1
- UL 60950-1
- EN 60950-1



NOTES

- M3 SCREW USED TO SECURE UNITS BASEPLATE TO OTHER SURFACES SUCH AS HEATSINK MUST NOT EXCEED 0.100"
 (2.54mm) DEPTH BELOW THE SURFACE OF THE BASEPLATE
- 2) APPLIED TORQUE PER SCREW SHOULD NOT EXCEED 6in-lb (0.7Nm)
- 3) PINS 4-5 ARE 0.062" (1.57mm) DIA. WITH 0.100" (2.54mm) DIA. STANDOFF SHOULDERS
- 4) PINS 1-3 ARE 0.040"(1.02mm) DIA. WITH 0.080" (2.03mm) DIA. STANDOFF SHOULDERS
- 5) ALL PINS: MATERIAL: COPPERY ALLOY FINISH: MATTE TIN OVER NICKEL PLATE
- 6) OTHER PIN EXTENSION LENGTHS AVAILABLE
- 7) BASEPLATE FLATNESS TOLERANCE IS 0.004" (0.10mm)TIR FOR SURFACE
- 8) UNDIMENSIONED COMPONENTS ARE SHOWN FOR VISUAL REFERENCE ONLY
- 9) ALL DIMENSIONS IN INCHES(mm) TOLERANCES: X.XX IN +/-0.02 (X.Xmm +/-0.5mm)
- 10) Weight: 1.62 oz (46.2 g)

PIN DESIGNATIONS

Pin	Name	Function
1	Vin(+)	Positive input voltage
2	ON/OFF	Logic control input to turn converter on/off.
3	Vin(-)	Negative input
4	Vout(-)	Negative output
5	Vout(+)	Positive output

Doc.# 005-0006514 Rev. H

SynCor® Technical Specification

Input:36-75V Output:12V Current:25A Part No.:SQ60120EPB25

SQ60120EPB25 Electrical Characteristics

Ta = 25 °C, airflow rate = 300 LFM, Vin = 48 V dc unless otherwise noted; full operating temperature range is -40 °C to +100 °C baseplate temperature with appropriate power derating. Specifications subject to change without notice.

Parameter	Min.	Typ.	Max.	Units	Notes & Conditions
ABSOLUTE MAXIMUM RATINGS					
Input Voltage					
Non-Operating	-1		100	V	Continuous
Operating	36	48	75	V	Continuous
Operating Transient Protection			100	V	100 ms transient, square wave
Isolation Voltage					
Input to Output			2250	V	
Operating Temperature	-40		100	°C	
Storage Temperature	-45		125	°C	
Voltage at ON/OFF input pin	-2		18	V	
INPUT CHARACTERISTICS		1			
Operating Input Voltage Range	36	48	75	V	100 ms transient, square wave
Input Under-Voltage Lockout					
Turn-On Voltage Threshold	31		34	V	
Turn-Off Voltage Threshold	28		32	V	
Lockout Voltage Hysteresis	2		4	V	
Input Over-Voltage Shutdown				V	
Turn-On Voltage Threshold		95		V	Vin when unit turns on after shutdown event
Turn-Off Voltage Threshold		105		V	Vin when unit will shut down
Maximum Input Current			9.5	A	100% Load, 36 Vin
No-Load Input Current			130	mA	,
Disabled Input Current			8.5	mA	
Input Reflected-Ripple Current		7.5	14	mA	RMS through 10µH inductor
Input Terminal-Ripple Current		200	350	mA	RMS, full load
Recommended Input Fuse (see Note 1)			20	A	Fast blow external fuse recommended
Recommended External Input Capacitance		47		μF	Typical ESR 0.1-0.2 Ω
Input Filter Component Values (L\C)		1.0\9.4		μH\μF	Internal values
OUTPUT CHARACTERISTICS		- (-		P W	
Output Voltage Set Point	12.2	12.25	12.3	V	Vin = 65 V, no load, Ta = 25 °C
Output Voltage Regulation					
Over Line		+0.05\-1.0		V	
Over Load		+0\-1.2		V	
Over Temperature		±300		mV	
Total Output Voltage Range	10.45		12.65	V	Over sample, line, load, temperature & life
Output Voltage Ripple and Noise					20 MHz bandwidth; see Note 2
		90	170	mV	Full load
Peak-to-Peak					
Peak-to-Peak RMS			-		
	0	30	65 25	mV A	Full load Subject to thermal derating
RMS Operating Output Current Range	0		65	mV A	Full load
RMS Operating Output Current Range Output DC Current-Limit Inception	0	30	65	mV	Full load Subject to thermal derating
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage	0	30 35	65	mV A A V	Full load Subject to thermal derating Output Voltage 10% Low
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled	0	30 35 10.5	65 25	mV A A	Full load Subject to thermal derating
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance	0	30 35 10.5	65	mV A A V mA	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance DYNAMIC CHARACTERISTICS	0	30 35 10.5	65 25	mV A A V mA	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance DYNAMIC CHARACTERISTICS Output Voltage during Load Current Transient	0	30 35 10.5 50	65 25 3,000	mV A V mA µF	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output 12Vout at 12A Resistive Load
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance DYNAMIC CHARACTERISTICS Output Voltage during Load Current Transient Step Change in Output Current (0.1 A/μs)	0	30 35 10.5 50 1800	65 25 3,000 2500	mV A V mA µF mV	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output 12Vout at 12A Resistive Load 50% to 75% to 50% Iout max
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance DYNAMIC CHARACTERISTICS Output Voltage during Load Current Transient Step Change in Output Current (0.1 A/µs) Settling Time		30 35 10.5 50	65 25 3,000	mV A V mA µF	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output 12Vout at 12A Resistive Load
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance DYNAMIC CHARACTERISTICS Output Voltage during Load Current Transient Step Change in Output Current (0.1 A/µs) Settling Time Turn-On Transient	0	30 35 10.5 50 1800 200	65 25 3,000 2500 300	mV A V mA μF mV μs	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output 12Vout at 12A Resistive Load 50% to 75% to 50% Iout max To within 1% Vout nom
RMS Operating Output Current Range Output DC Current-Limit Inception Output DC Current-Limit Shutdown Voltage Back-Drive Current Limit while Disabled Maximum Output Capacitance DYNAMIC CHARACTERISTICS Output Voltage during Load Current Transient Step Change in Output Current (0.1 A/µs) Settling Time		30 35 10.5 50 1800	65 25 3,000 2500	mV A V mA µF mV	Full load Subject to thermal derating Output Voltage 10% Low Negative current drawn from output 12Vout at 12A Resistive Load 50% to 75% to 50% Iout max

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SynCor[®] Technical Specification

Input:36-75V Output:12V Current:25A Part No.:SQ60120EPB25

SQ60120EPB25 Electrical Characteristics (continued)

Ta = 25 °C, airflow rate = 300 LFM, Vin = 48 V dc unless otherwise noted; full operating temperature range is -40 °C to +100 °C baseplate temperature with appropriate power derating. Specifications subject to change without notice.

Parameter	Min.	Тур.	Max.	Units	Notes & Conditions
ISOLATION CHARACTERISTICS			ľ		
Isolation Voltage (dielectric strength)			2250	V	See Absolute Maximum Ratings
Isolation Resistance		30		MΩ	
Isolation Capacitance (input to output)		N/A		pF	See Note 3
TEMPERATURE LIMITS FOR POWER DERATIN	IG CURVES				·
Semiconductor Junction Temperature			125	°C	Package rated to 150 °C
Board Temperature			125	°C	UL rated max operating temp 130 °C
Transformer Temperature			135	°C	
EFFICIENCY					
100% Load		95		%	
50% Load		95.5		%	
FEATURE CHARACTERISTICS					
Switching Frequency	220	250	280	kHz	
ON/OFF Control (Option P)					
Off-State Voltage	-1		0.8	V	
On-State Voltage	2.4		18	V	
ON/OFF Control (Option N)					
On-State Voltage	-1		0.8	V	
Off-State Voltage	2.4		18	V	
ON/OFF Control (Either Option)					Application notes Figure B
Pull-Up Voltage		5		V	
Pull-Up Resistance		124		kΩ	
Output Over-Voltage Protection		16		V	Over full temp range; no load
Over-Temperature Shutdown OTP Trip Point	140		150	°C	Average PCB Temperature
Over-Temperature Shutdown Restart Hysteresis		10		°C	
RELIABILITY CHARACTERISTICS					
Calculated MTBF (Telcordia) TR-NWT-000332		2.6		106 Hrs.	80% load, 200LFM, 40 °C Ta
Calculated MTBF (MIL-217) MIL-HDBK-217F		2.1		10 ⁶ Hrs.	80% load, 200LFM, 40 °C Ta

Note 1: Safety certification tests were carried out using 20A fast blow fuse. Fuse interruption characteristics have to be taken into account while designing input traces. User should ensure that Input trace is capable of withstanding fault currents

Note 2: For applications requiring reduced output voltage ripple and noise, consult SynQor applications support (e-mail: support@synqor.com) Note 3: Isolation capacitance can be added external to the module.

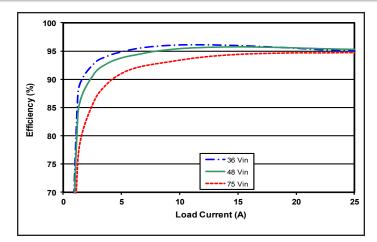


Standards Compliance & Qualification Testing

Notes & Conditions

Note: An external input fuse must always be used to meet these safety requirements. Contact SynQor for official safety certificates on new releases or download from the SynQor website.

Parameter	# Units	Test Conditions
QUALIFICATION TESTING		
Life Test	32	95% rated Vin and load, units at derating point, 1000 hours
Vibration	5	10-55 Hz sweep, 0.060" total excursion, 1 min./sweep, 120 sweeps for 3 axis
Mechanical Shock	5	100 g minimum, 2 drops in x, y and z axis
Temperature Cycling	10	-40 °C to 100 °C, unit temp. ramp 15 °C/min., 500 cycles
Power/Thermal Cycling	5	Toperating = min to max, Vin = min to max, full load, 100 cycles
Design Marginality	5	Tmin-10 °C to Tmax+10 °C, 5 °C steps, Vin = min to max, 0-105% load
Humidity	5	85 °C, 85% RH, 1000 hours, continuous Vin applied except 5 min/day
Solderability	15 pins	MIL-STD-883, method 2003



Technical Figures

Figure 1: Efficiency at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

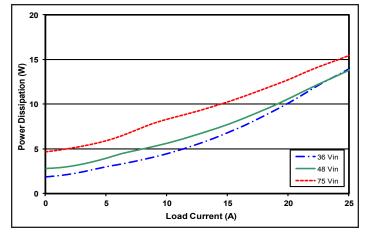


Figure 3: Power dissipation at nominal output voltage vs. load current for minimum, nominal, and maximum input voltage at 25°C.

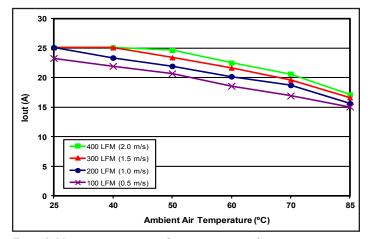


Figure 5: Maximum output current derating curve vs. ambient air temperature (with baseplate only) at different airflow rates with air flowing from Vin- to Vin+ (nominal input voltage).

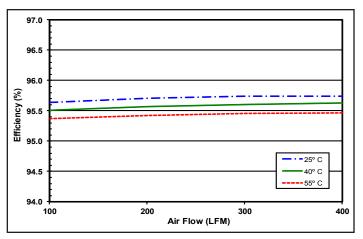


Figure 2: Efficiency at nominal output voltage and 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C (nominal input voltage).

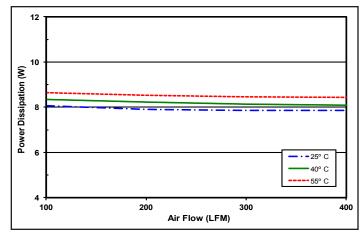


Figure 4: Power dissipation at nominal output voltage and 60% rated power vs. airflow rate for ambient air temperatures of 25°C, 40°C, and 55°C (nominal input voltage).

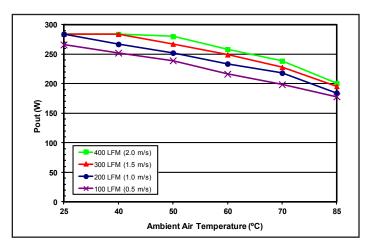
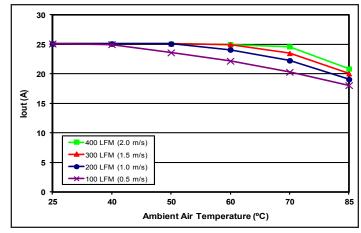


Figure 6: Maximum output power derating curve vs. ambient air temperature (with baseplate only) at different airflow rates with air flowing from Vin- to Vin+ (nominal input voltage).



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Technical Figures

Figure 7: Maximum output current derating curve vs. ambient air temperature (with both baseplate and 0.5" heat sink) at different airflow rates with air flowing from Vin- to Vin+ (nominal input voltage).

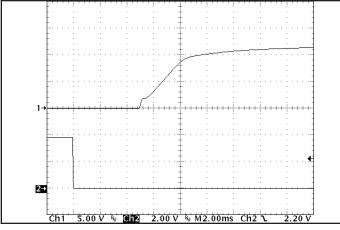


Figure 9: Turn-on transient at half load (resistive load) and 3mF output capacitance (2 ms/div). Input voltage pre-applied. Ch 1: Vout (5V/div). Ch 2: ON/OFF input (2V/div).

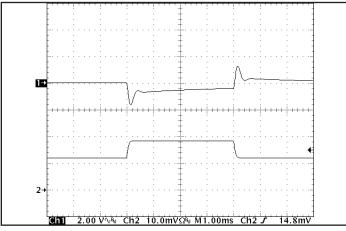
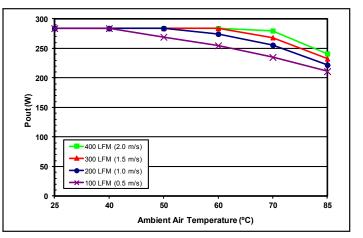
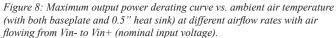


Figure 11: Output voltage response to step-change in load current (50%-75%-50% of lout(max); $dI/dt = 0.1A/\mu s$). Load cap: $100\mu F$, electrolytic cap, 15uFtantalum cap, and $1\mu F$ ceramic cap. Ch 1: Vout (2V/div), Ch 2: lout (10A/div).





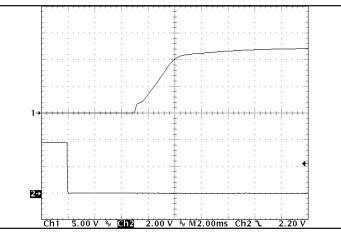


Figure 10: Turn-on transient at zero load and 3mF output capacitance (2ms/div). Input voltage pre-applied. Ch 1: Vout (5V/div). Ch 2: ON/OFF input (2V/div).

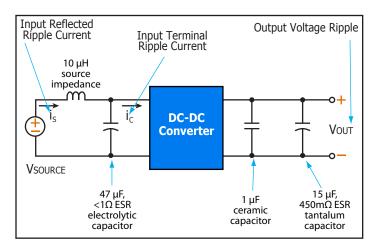


Figure 12: Test set-up diagram showing measurement pints for Input Terminal Ripple Current (Figure 13), Input Reflected Ripple current (Figure 14) and Output Voltage Ripple (Figure 15).

School (1975) Technical Figures

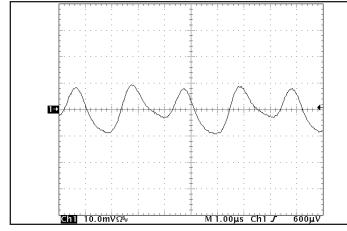


Figure 13: Input Terminal Ripple Current, ic, at full rated output current and nominal input voltage with 10 μ H source impedance and 47 μ F electrolytic capacitor (200 mA/div). See Figure 12.

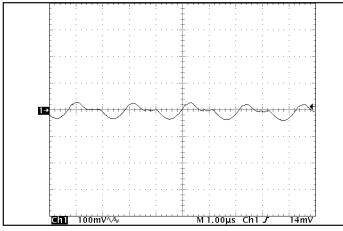


Figure 15: Output voltage ripple at nominal input voltage and rated load current (100 mV/div). Load capacitance: $1\mu F$ ceramic capacitor and $15\mu F$ tantalum capacitor. Bandwidth: 20 MHz. See Figure 12.

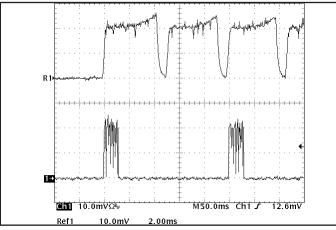


Figure 17: Load current (20 A/div) as a function of time when the converter attempts to turn on into a 1 m Ω short circuit. Top trace 2 ms/div) is an expansion of the on-time portion of the bottom trace.

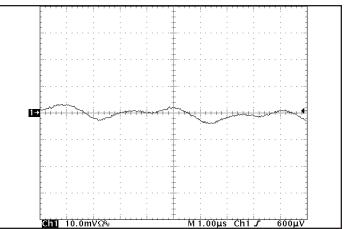


Figure 14: Input reflected ripple current, is, through a 10 μ H source inductor, using a 47 μ F electrolytic input capacitor (10 mA/div). See Figure 12.

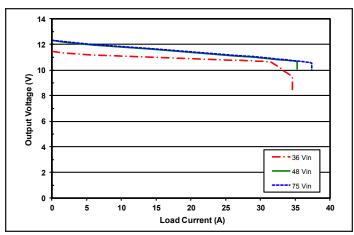


Figure 16: Output voltage vs. load current showing typical current limit curves and converter shutdown points.

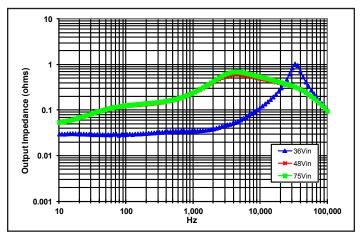
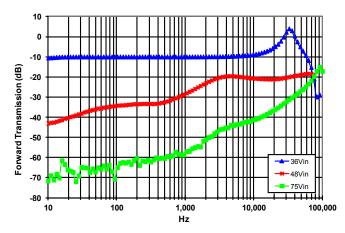


Figure 18: Magnitude of incremental output impedance (Zout = Vout/Iout) for minimum, nominal, and maximum input voltage at full rated power.





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Technical Figures

Figure 19: Magnitude of incremental forward transmission (FT = Vout/Vin) for minimum, nominal, and maximum input voltage at full rated power.

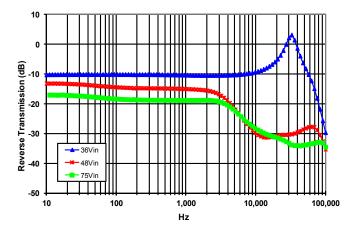


Figure 20: Magnitude of incremental reverse transmission (RT = Iin/Iout) for minimum, nominal, and maximum input voltage at full rated power.

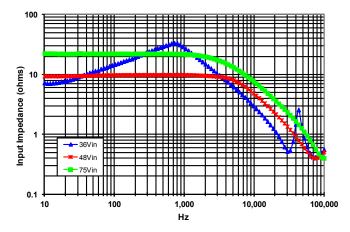


Figure 21: Magnitude of incremental input impedance (Zin = Vin/lin) for minimum, nominal, and maximum input voltage at full rated power.

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BASIC OPERATION AND FEATURES

With voltages dropping and currents rising, the economics of an Intermediate Bus Architecture (IBA) are becoming more attractive, especially in systems requiring multiple low voltages. IBA systems separate the role of isolation and voltage scaling from regulation and sensing. The BusQor series bus converter provides isolation and a coarse voltage step down in one compact module, leaving regulation to simpler, less expensive non-isolated converters.

In Figure A below, the BusQor module provides the isolation stage of the IBA system. The isolated bus then distributes power to the non-isolated buck regulators to generate the required voltage levels at the point of load. In this case, the bucks are represented with SynQor's NiQor series of nonisolated dc-dc converters. In many applications requiring multiple low voltage outputs, significant savings can be achieved in board space and overall system costs.

When designing an IBA system with bus converters, the designer can select from a variety of bus voltages. While there is no universally ideal bus voltage, most designs employ one of the following: 12V, 9V, 7.5V, 5V, or 3.3V. Higher bus voltages can lead to lower efficiency for the buck regulators but are more efficient for the bus converter and provide lower board level distribution current. Lower bus voltages offer the opposite trade offs.

SynQor's SQ BusQor modules accept the full 48/60V telecom bus voltage and have a narrow output voltage range compatible with many non-isolated point of load converters. When used in IBA systems, the output variation of the BusQor must be in accordance with the input voltage range of the non-isolated converters being employed.

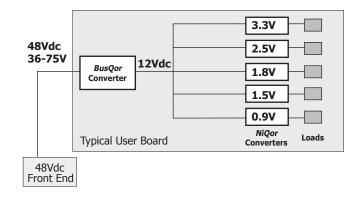
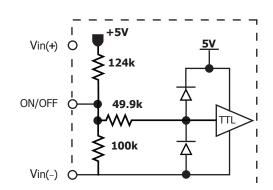


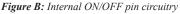
Figure A: Example of Intermediate Bus Architecture using BusQor bus converter and NiQor non-isolated converters

CONTROL FEATURES

REMOTE ON/OFF (Pin 2): The ON/OFF input, Pin 2, permits the user to control when the converter is on or off. This input is referenced to the return terminal of the input bus, Vin(-). There are two versions of the converter that differ by the sense of the logic used for the ON/OFF input.

In the positive logic version, the ON/OFF input is active high (meaning that a high turns the converter on). In the negative logic version, the ON/OFF signal is active low (meaning that a low turns the converter on). Figure B is a detailed look of the internal ON/OFF circuitry.







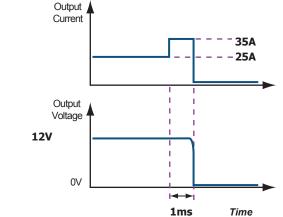
Protection Features

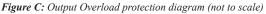
Input Under-Voltage Lockout: The converter is designed to turn off when the input voltage is too low, helping avoid an input system instability problem, described in more detail in the application note titled "Input System Instability" available on www.SynQor.com. The lockout circuitry is a comparator with DC hysteresis. When the input voltage is rising, it must exceed the typical Turn-On Voltage Threshold value (listed on the specification page) before the converter will turn on. Once the converter is on, the input voltage must fall below the typical Turn-Off Voltage Threshold value before the converter will turn off. Also see Fig E.

Output Current Limit: The output of the BusQor module is electronically protected against output overloads. When an overload current greater than the "DC Current-Limit Inception" specification is drawn from the output, the output of the BusQor will shutdown to zero volts after the 1ms have elapsed (see Fig C). The shutdown period lasts for a typical period of 200ms (Fig D) after which the BusQor tries to power up again. If the overload persists, the output voltage will go through repeated cycles of shutdown and restart with a duty cycle of 20ms (On) and 200ms (Off) respectively. The BusQor module returns (auto resetting) to normal operation once the overload is removed. The BusQor is designed to survive in this mode indefinitely without damage and without human intervention. **Output Short Circuit Protection:** When the output of the BusQor module is shorted, a peak current of typically 35A will flow into the short circuit for a time of about 1ms. The output of the BusQor will shutdown to zero volts after the 1ms have elapsed (Fig C & D). The shutdown period lasts for a time of 200ms, at the end of which the BusQor module tries to power up again. If the short circuit persists, the output voltage will go through repeated cycles of shutdown and restart with a duty cycle of 20ms (On) and 200ms (Off) respectively. The BusQor module returns (auto resetting) to normal operation once the short circuit is removed. The BusQor is designed to survive in this mode indefinitely without damage and without human intervention.

In the Auto resetting mode, also referred to as "Hiccup" mode, the power drawn from the 48V input is about 5W, most of which is dissipated into the external fault. It is important that copper traces and pads from the output circuit be designed to withstand the short term peaks, although the average current into the fault may be as low as 50mA typical. See Fig 17 for appropriate waveform.

Over-Temperature Shutdown: A temperature sensor on the converter senses the average temperature of the module. The thermal shutdown circuit is designed to turn the converter off when the temperature at the sensed location reaches the Over-Temperature Shutdown value. It will allow the converter to turn on again when the temperature of the sensed location falls by the amount of the Over-Temperature Shutdown Restart Hysteresis value.





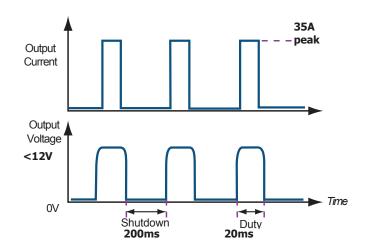


Figure D: Output Short Circuit and Auto-Resetting protection diagram (not to scale)



APPLICATION CONSIDERATIONS

Startup Inhibit Period: Figure E details the Start-Up Inhibit Period for the BusQor module when Vin is applied with the On/Off pin asserted (enabled). Before time t_1 , when the input voltage Off pin asserted (enabled). Before time t_1 , when the input voltage is below the UVL threshold, the unit is disabled by the Input Under-Voltage Lockout feature. When the input voltage rises above the UVL threshold, the Input Under-Voltage Lockout is released, and a typical Startup Inhibit Period of 8ms from a Vin start-up. Once the module recognizes it should be on it waits an additional 4ms before it starts. When starting from an on/ off event the module has an inhibit time of 200ms like our standard products. The output builds up to 90% or higher of the nominal voltage and stabilizes to its nominal output voltage in a total time of 15ms typical.

At time t_2 , when the On/Off pin is de-asserted (disabled), the BusQor module stops switching immediately. Fall time from 12V to 0V is dependent on output capacitance load current, and any parasitic trace inductance in the output load circuit.

At time t_{3} , when the On/Off pin is re-asserted (enabled), the BusQor module output begins to build up after the inhibit period of 200ms typical.

Refer to the Control Features section of the data sheet for details on enabling and disabling methods for N and P logic type modules.

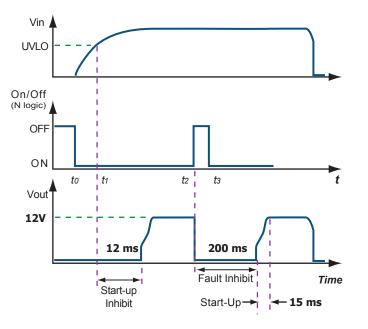


Figure E: Power Up/Down Diagram (not to scale) showing Start-Up Inhibit Period.



Droop Current Sharing:

This product has primary side regulation only. Primary side regulation results in a voltage droop as the load current increase (droop characteristic). The inherent impedance of the power stage due to its droop characteristic can be used to balance the load current between multiple modules that are placed in parallel. The graphs in this section show two units by way of example, but there is no fundamental limit to the number of units that can be placed in parallel.

When multiple units having droop characteristics are placed in parallel, the current sharing accuracy is determined by the output voltage set point accuracy relative to the total output voltage droop for a given load current range. A difference in voltage between two units will cause a differential current to flow out of one unit and into the other. Figure F shows an example with two units with an output voltage mismatched by 0.5%. In this example, when Unit A is at 100% of its full rated load current, Unit B is only at 90%, effectively reducing the total available current by 5%. SynQor uses factory calibration of each unit to ensure that output voltage is well matched. **Temperature Mismatch Self Balancing:** the slope of the output voltage droop characteristic of semi regulated module increases with increased temperature. Therefore, if a paralleled unit were hotter than its neighbor because it is supplying more load current, then its output voltage will droop further forcing the cooler module(s) to increase its load current. The increase in effective output resistance due to an increase in temperature acts as a self correcting mechanism (Fig G) when two or more modules are connected in a parallel configuration.

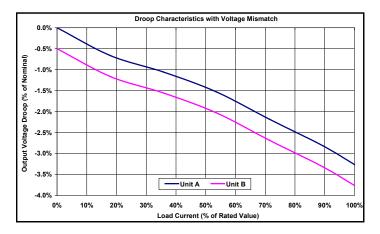


Figure F: Example of Droop Characteristic with Output Voltage Mismatch.

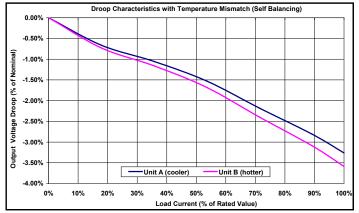
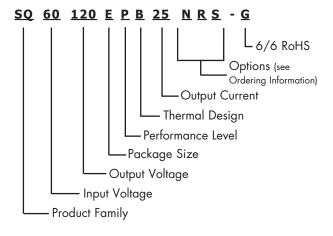


Figure G: Droop Characteristic with Temperature Mismatch (self balancing)



Part Numbering System

The part numbering system for SynQor's dc-dc converters follows the format shown in the example below.



The first 12 characters comprise the base part number and the last 3 characters indicate available options. The "-G" suffix indicates 6/6 RoHS compliance.

Application Notes

A variety of application notes and technical white papers can be downloaded in pdf format from our website.

RoHS Compliance: The EU led RoHS (Restriction of Hazardous Substances) Directive bans the use of Lead, Cadmium, Hexavalent Chromium, Mercury, Polybrominated Biphenyls (PBB), and Polybrominated Diphenyl Ether (PBDE) in Electrical and Electronic Equipment. This SynQor product is 6/6 RoHS compliant. For more information please refer to SynQor's RoHS addendum available at our RoHS Compliance / Lead Free Initiative web page or e-mail us at rohs@synqor.com.

Ordering Information

The tables below show the valid model numbers and ordering options for converters in this product family. When ordering SynQor converters, please ensure that you use the complete 15 character part number consisting of the 12 character base part number and the additional characters for options. Add "-G" to the model number for 6/6 RoHS compliance.

Model Number	Input Voltage	Output Voltage	Max Output Current
SQ60120EPB25xyz	36-75V	12V	25A

The following options must be included in place of the $w \times y z$ spaces in the model numbers listed above.

Not all combinations make valid part numbers, please contact SynQor for availability.

Options Description:						
Thermal Design	Enable Logic	Pin Style	Feature Set			
B - Baseplated	N - Negative P - Positive	K - 0.110" N - 0.145" R - 0.180" Y - 0.250"	S - Standard			

Contact SynQor for further information and to order:

 Phone:
 978-849-0600 Toll Free: 888-567-9596
 Fax:
 978-849-0602

 E-mail:
 power@synqor.com
 Web:
 www.synqor.com

 Address:
 155 Swanson Road, Boxborough, MA 01719
 USA

Warranty

SynQor offers a three (3) year limited warranty. Complete warranty information is listed on our website or is available upon request from SynQor.

PATENTS

7,050,309

8.149.597

SynOor's patents include the following:

7,765,687

8,644,027

Doc.# 005-0006514 Rev. H

7,787,261

SynQor holds numerous U.S. patents, one or more of which apply to most of its power conversion products. Any that apply to the product(s) listed in this document are identified by markings on

the product(s) or on internal components of the product(s) in accordance with U.S. patent laws.